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Koninklijke Philips Electronics N.V.
Groenewoudseweg 1
5621 BA Eindhoven
PAYS-BAS

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Device and method for determining the current flowing through a gas discharge lamp

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Device and method for determining the current flowing through a gas discharge lamp
EPO - DG 1

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The present invention relates to a device and method for determining the current flowing through a gas discharge lamp. The present invention also relates to an electronic ballast for operating a gas discharge lamp.

Nowadays power control devices or ballasts are widely used for controlling the power supplied to discharge lamps such as fluorescent lamps. Ballasts can for example be employed to optimize the preheating and ignition of the discharge lamp, to maintain a constant power to the electric discharge lamp for the purpose of maintaining a selected light intensity or for the purpose of controlled dimming to a fixed, but adjustable, power level of the discharge lamp.

Modern electronic ballasts comprise a switched-mode power supply (SMPS) connected between the supply voltage (typically the mains) and the discharge lamp. In a three stage ballast circuit the first stage of the switched-mode power supply comprises a preconditioner, for example a double rectifier for rectifying the mains (230 V, 50 Hz, 1 phase), combined with an up-converter. The second stage may comprise a down-converter (DC-DC converter), also called a forward or buck converter, for stabilizing the output current. The third stage of the ballast circuit comprises a commutator bridge and ignitor to implement a square wave current operation of the lamp. In a two-stage ballast topology the down-converter and commutator bridge are replaced by a half-bridge commutating forward (HBCF) or a full-bridge commutating forward (FBCF) topology.

The half-bridge commutating forward (HBCF) circuit corresponds to a full-bridge commutating forward (FBCF) circuit wherein a part of the bridge is replaced by two (electrolytic) bridge capacitors in series. The ballast in this topology comprises an up-converter in combination with a half bridge acting as double down-converter. This two stage ballast topology for operating a HID lamp is relatively simple and relatively inexpensive.

The control of the power supplied to the lamp may be based on the outcome of measurements of various lamp parameters, such as the actual current flowing through the commutating forward coil element (HBCF-coil or FBCF-coil). This converter current may be used as a measure of the actual current flowing through the lamp. Measurement of the

HBCF/FBCF coil current or converter current may be implemented in various ways, each of them exhibiting a number of drawbacks.

One of the methods of determining the converter current is to provide a current shunt, for example by connecting a sense resistor in series with the HBCF coil. Using a differential amplifier the differential voltage across the sense resistor is measured. Based on the known resistance value of the sense resistor the current flowing through the sense resistor and consequently the actual converter current, i.e. the current flowing through the HBCF coil, can be determined. However, one of the disadvantages of this method is that a high specification operational amplifier, that is an operational amplifier with a large common mode rejection ratio, is needed, resulting in considerable costs of the ballast. Furthermore, the signals measured with the amplifier have relatively small values, because the resistance value of the sense resistor should be as small as possible to minimize the losses induced by the insertion of the sense resistor. These small values lead to a poor signal to noise ratio.

A further method of determining the converter current is to use a current transformer, for example by connecting the primary windings of a current transformer in series with the HBCF coil. The secondary windings of the transformer will then provide a signal proportional to the converter current. However, one of the disadvantages of this method is that not only the high frequency components of the current signal, but also the low frequency components of the current signal are to be transferred. To guarantee the transfer of the low frequency components, cf. the low frequent commutation signal, of the current signal, a relatively bulky transformer is needed.

Furthermore, asymmetrical current operation of the discharge lamp cannot be detected. During the start phase and/or at the end of the life (EOL) of the lamp, the lamp behaviour may be irregular causing an asymmetrical lamp load of the above mentioned commutating forward circuit. The lamp, for example, may be conducting in one half period of the duty cycle of the switched-mode power supply, and may be non-conducting in the other half period. In the above mentioned full bridge commutating forward circuit the resulting DC component cannot be determined. In the above mentioned half bridge commutating forward circuit the asymmetrical load of the half bridge results in a displacement of the midpoint voltage in the bridge capacitor series circuit, i.e. the voltage that is the voltage at the junction between the first and second bridge capacitors is increased or decreased. As a result of the above voltage drift, the maximum voltage rating of one of the bridge capacitors may be exceeded causing damage to the ballast.

It is an object of the invention to provide a device and method for determining the converter current and to provide an electronic ballast wherein the above mentioned drawbacks are obviated.

According to a first aspect of the invention this object is achieved in a device
5 for determining the current supplied to a discharge lamp by a commutating forward converter, the converter being connectable to a rail line for supplying a rail voltage and the converter comprising a first switching element, a second switching element and an output node between said switching elements for supplying said current to the discharge lamp, the device comprising a first current sense circuit for sensing the current on a first position
10 between the rail and the output node and a second current sense circuit for sensing the current on a second position between the output node and ground. By sensing the current on two positions, at a position in the upper half of the bridge and at a position in the lower half of the bridge, only the high frequency component of the current signal (typically in the range of 30 kHz - 250 kHz) is to be determined. Sensing the low frequency components, such as the
15 commutation frequency, typically in the range of 50-400 Hz, can be dispensed with. This enables the use of relatively small current transformers, the total volume of which is less than for the above discussed single current transformer method. Furthermore, for half bridge applications asymmetrical operation can be detected. This enables control circuitry of the power supply to adapt the duty cycle of the switching elements so as to correct the midpoint
20 voltage and hence the voltages across the bridge capacitors to safe values. Also for full bridge applications a DC component can be detected.

Furthermore the power losses with this measurement method are reduced as compared to the losses arising in the above discussed current shunt method. Also the output signals need no further amplification, which avoids noise or interference problems.

25 In a preferred embodiment the first sense circuit comprises a first current transformer including a primary winding connected on said first position and the second sense circuit comprises a second current transformer including a primary winding connected on said second position, the secondary windings of the first and second current transformers being connected in series for providing a combined signal representative of the converter
30 current.

According to another aspect of the present invention an electronic ballast is provided for operating a gas discharge lamp, comprising:

- a switched mode power supply (SMPS) circuit for supplying power to the discharge lamp, the switched mode power supply circuit comprising a half or full bridge

commutating forward converter including at least a rail line for supplying a rail voltage, a first switching element, a second switching element and an output node between said switching elements for supplying current to the lamp;

- a current determining circuit for providing a signal representative of the
5 converter current;

wherein the current determining circuit comprise a first current sense circuit for sensing the current on a first position between the rail and the output node and a second current sense
circuit for sensing the current on a second position between the output node and ground.

- In a preferred embodiment the ballast comprises a gate driving circuit,
10 connected to the gates of the first switching element and the second switching element, and to the current determining circuit for controlling the switching of the switching elements based on said signal representative of the converter current. The signal representative of the converter current is fed back to the control circuitry that controls the duty cycle of the switching elements of the switched-mode power supply. Based on this signal the duty cycle
15 of the switching elements may be adapted by the control circuitry.

- According to still another aspect of the present invention a method is provided of determining the current supplied by a commutating forward converter to a gas discharge lamp, the converter including at least a rail line for supplying a rail voltage, a first switching element, a second switching element and an output node between said switching elements for
20 supplying current to the lamp, the method comprising the steps of:

- sensing the current in the converter on a first position between the rail line and the output node and providing a first output signal;
- sensing the current in the converter on a second position between the output
node and ground and providing a second output signal;
- 25 adding the first and second output signals to providing a third output signal representative of the converter current.

- When the first signal is the current measured on the first position, the second signal is the current measured on the second position and the third signal is the summation of the current measured on the first position and the simultaneously measured current on the
30 second position, a measure is obtained of the current flowing through the HBCF coil. This current is a measure of the current flowing through the lamp.

Further advantages, features and details of the present invention will be elucidated with reference to the annexed drawings, in which:

Fig. 1 shows a schematic circuit diagram of an electronic ballast according to a first preferred embodiment of the present invention;

5 Fig. 2 shows a graph of the current signal of the upper switching element, the current signal of the lower switching element and the converter current;

Fig. 3 shows a graph of the combined current signals of the upper and lower switching element and the converter current; and

10 Fig. 4 shows a schematic circuit diagram of an electronic ballast according to a second preferred embodiment of the present invention.

Figure 1 shows a two-stage ballast for a high intensity discharge lamp (LA). The first stage (I) of the ballast comprises a rectifier 2 for converting the AC supply voltage (typically a 230 V 50 Hz mains) to a DC supply voltage and an up-converter or boost-
15 converter 3 for boosting the DC supply voltage. In figure 1 a typical topology of a boost-converter or up-converter is shown. The boost-converter inter alia is composed of an inductor (Lboost), a switching element (T) and a diode (D).

The second stage (II) of the ballast as shown in figure 1 comprises a half
20 bridge commutating forward (HBCF) circuit acting as a double down-converter. The HBCF circuit includes a first MOSFET T1, a second MOSFET T2, a first and a second (internal) body diode D1 and D2, an inductor Lhbcf in series with the lamp, a lamp capacitor Cr connected parallel to the lamp, and two electrolytic bridge capacitors Cs1 and Cs2 connected in series. The half bridge commutating forward circuit is operated in the critical
25 discontinuous mode to allow zero voltage switching. Each half commutation period (commutation frequency in the order of 100 Hz), one MOSFET (the first MOSFET T1 or the second MOSFET T2) is operated in combination with the diode (D2 or D1) of the other MOSFET. Switching of the MOSFETS is accomplished by a duty cycle control circuit, as is schematically shown in figure 1. This circuit controls the duty cycle of the half bridge
30 commutating forward circuit. The control may be made dependent on the converter current or at least a signal representative of the converter current, as determined according to the invention.

The primary windings of a first current transformer CT1 are connected between the rail line and the first MOSFET T1. The current transformer can equally well be

connected between the MOSFET T1 and the output node (O) between both MOSFETS T1 and T2. The primary windings of a second current transformer CT2 are connected between the second MOSFET T2 and ground or between the output node (O) and the second MOSFET T2. The secondary windings of the first transformer and second transformer are
5 connected in series.

In figures 2 and 3 measurements of the current flowing through the core of first transformer CT1 in the upper part of the half bridge commutating forward circuit and the current flowing through the core of the second transformer CT2 in the lower part of the half bridge commutating forward circuit are displayed. Figure 2 in fact shows three signals. .
10 Signal A represents the response in time of the first current transformer CT1 belonging to the upper MOSFET (T1), while signal B is the response in time of the current transformer CT2 belonging to the lower MOSFET (T2). Signal C shows the actual converter current as function of time. The left part of the figure shows a first commutation half period, while the right part of the figure shows a subsequent half commutation period.

From figure 2 clearly follows that only the high frequency components of the currents through the transformer cores are transferred well and the low frequent (commutation) frequency components feed out quickly. In figure 3 signal D shows a signal that is equal to signal A added with signal B. It becomes clear that the effects of the weak low frequent response and the mean values are cancelled. The resulting current signal D gives
15 clear zero and peak current information. This information can be used to assess the operation of the converter current and consequently the lamp current. The resulting current signal as a consequence may be used to be sure of more pure AC lamp operation.

Although the low frequency part of the current signal feeds out quickly, the relatively small current transformers still show a small transfer of the low frequency part of the current signal, as can be derived from signals A and B in figure 2. Especially signal B
25 clearly shows that after commutation the zero level slowly approaches the "zero axis". After several high frequency periods the low frequency part of the signal has already disappeared. The above mentioned series connection of the secondary windings of the current transformers, resulting in signal D in figure 3, has the following advantages.

The slight low frequency transfers of both transformers will cancel out and will not influence the output signal (C). This implies that the low frequency transfer performance of the transformers has become less relevant). To improve the above mentioned cancellation of the low frequency components of signals A and B the transformers are chosen
30 such that the low frequency response of each transformer is substantially identical.

A further advantage is that the resulting signal D (figure 3) is unipolar or rectified. Regardless of the direction (positive or negative) the commutating forward converter sends the current (cf. figure 3, signal E), the maximum value of the current is positive. Consequently, the peak current detection circuit, which is connected to the secondary side of the current transformers, needs only to detect the positive maximum. Also in case of a zero-crossing the flank will always change from a negative flank, through a zero flank to a positive flank.

Figure 4 shows a two-stage ballast for a high intensity discharge lamp (LA), of which the first stage (I) corresponds to the ballast shown in figure 1. The second stage (II) of the ballast shows a full bridge commutating forward (FBCF) topology. The FBCF circuit includes a first MOSFET T1, a second MOSFET T2, a third MOSFET T3, and a fourth MOSFET T4, first, second, third and fourth (internal) body diodes D1-D4, a lamp inductor Lhbcf in series with the lamp, a lamp capacitor Cr connected parallel to the lamp, and one electrolytic capacitor Cs parallel to the second and third MOSFET. The full bridge commutating forward circuit is operated in the critical discontinuous mode to allow zero voltage switching. Similar to the ballast shown in figure 1, the primary windings of a first current transformer CT1 are connected between the rail line and the first MOSFET T1. The primary windings of a second current transformer CT2 are connected between the second MOSFET T2 and ground or between the output node (O) and the second MOSFET T2. The secondary windings of the first transformer and second transformer are connected in series. The combination of the signal derived from the first transformer and the signal derived from the second transformer will give a signal representative of the converter current.

Because during each high frequency period the actual peak current through the MOSFET is measured using the above mentioned current transformers, the current is always the same, both in the positive and in the negative period part of the low frequency current. A DC component (amplitude difference between the positive and negative low frequency period part) therefore is not possible.

The present invention is not limited to the above described preferred embodiments thereof; the rights sought are defined by the following claims, within the scope of which many modifications can be envisaged.

CLAIMS:

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1. Electronic ballast for operating a gas discharge lamp, comprising:

- a switched mode power supply (SMPS) circuit for supplying power to the discharge lamp, the switched mode power supply circuit comprising a half or full bridge commutating forward converter including at least a rail line for supplying a rail voltage, a first switching element (Q1), a second switching element (Q2) and an output node between said switching elements for supplying current to the lamp;

- a current determining circuit for providing a signal representative of the converter current;

wherein the current determining circuit comprise a first current sense circuit for sensing the current on a first position between the rail and the output node and a second current sense circuit for sensing the current on a second position between the output node and ground.

2. Electronic ballast according to claim 1, wherein the first sense circuit

comprises a first current transformer including a primary winding connected on said first

position and the second sense circuit comprises a second current transformer including a primary winding connected on said second position, the secondary windings of the first and second current transformers being connected in series for providing a combined signal representative of the converter current.

3. Electronic ballast according to claim 1 or 2, comprising a gate driving circuit, connected to the gates of the first switching element and the second switching element, and to the current determining circuit for controlling the switching of the switching elements based on said signal representative of the converter current.

4. Device for determining the current supplied by a commutating forward converter to a discharge lamp, the converter being connectable to a rail line for supplying a rail voltage and the converter comprising a first switching element, a second switching element and an output node between said switching elements for supplying said current to the discharge lamp, the device comprising a first current sense circuit for sensing the current on a

first position between the rail and the output node and a second current sense circuit for sensing the current on a second position between the output node and ground.

5. Device according to claim 4, wherein the first sense circuit comprises a first current transformer including a primary winding connected on said first position and the second sense circuit comprises a second current transformer including a primary winding connected on said second position, the secondary windings of the first and second current transformers being connected in series for providing a combined signal representative of the converter current.

6. Method of determining the current supplied by a commutating forward converter to a gas discharge lamp, the converter including at least a rail line for supplying a rail voltage, a first switching element, a second switching element and an output node between said switching elements for supplying current to the lamp, the method comprising the steps of:

sensing the current in the converter on a first position between the rail line and the output node and providing a first output signal;

sensing the current in the converter on a second position between the output node and ground and providing a second output signal;

adding the first and second output signals to providing a third output signal representative of the converter current.

7. Method according to claim 6, wherein the first signal is the current measured on the first position, the second signal is the current measured on the second position and the third signal is the summation of the current measured on the first position and the simultaneously measured current on the second position.

8. Method according to any of claims 6 or 7, wherein the electronic ballast according to any of claims 1-3 and/or the device according to any of claims 4-5 is applied.

ABSTRACT:

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The present invention relates to a device for operating a gas discharge lamp, comprising a switched mode power supply circuit for supplying power to the discharge lamp, the switched mode power supply circuit comprising a half or full bridge commutating forward converter including at least a rail line for supplying a rail voltage, a first switching
5 element, a second switching element and an output node between said switching elements for supplying current to the lamp, a current determining circuit for providing a signal representative of the converter current, wherein the current determining circuit comprise a first current sense circuit for sensing the current on a first position between the rail and the
10 output node and a second current sense circuit for sensing the current on a second position between the output node and ground. The invention also relates to a method and an electronic ballast for operating a gas discharge lamp.

Fig. 1

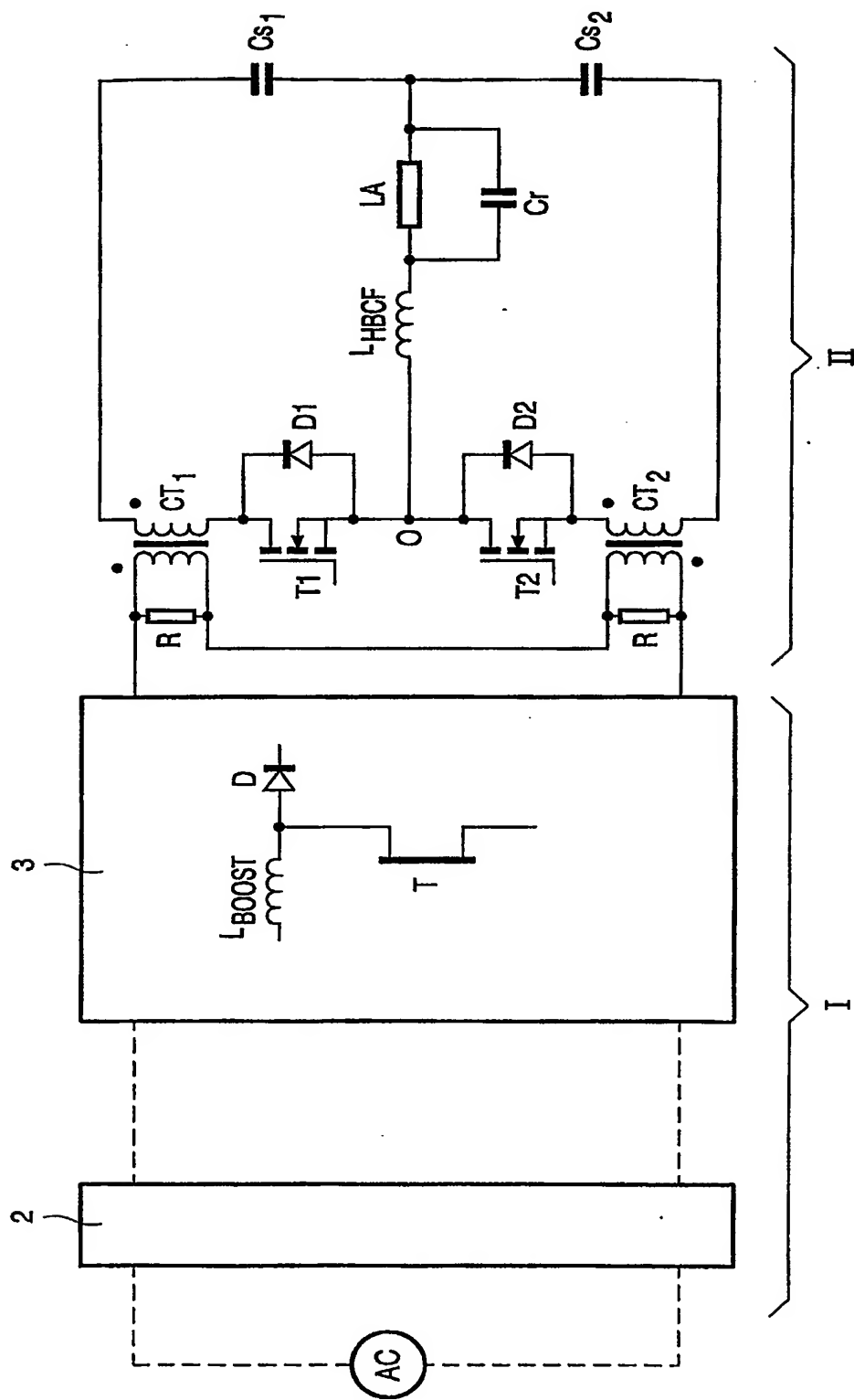


FIG. 1

2/3

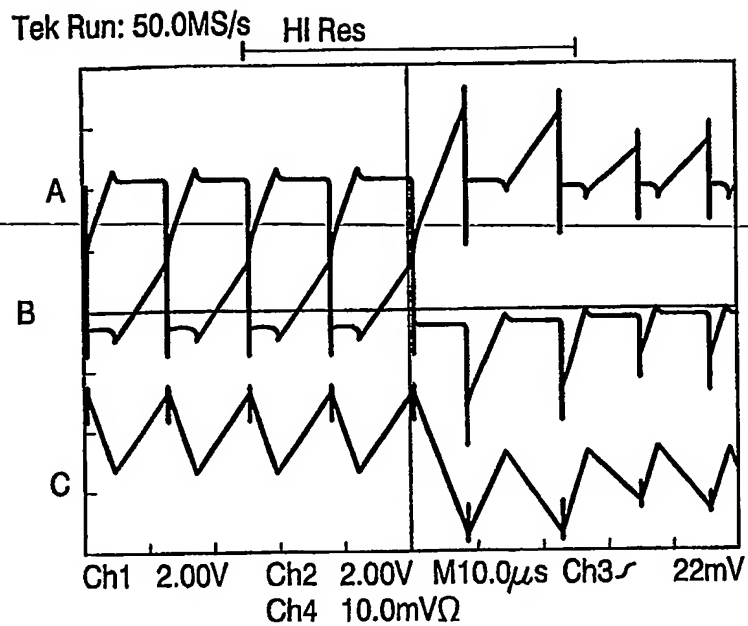


FIG. 2

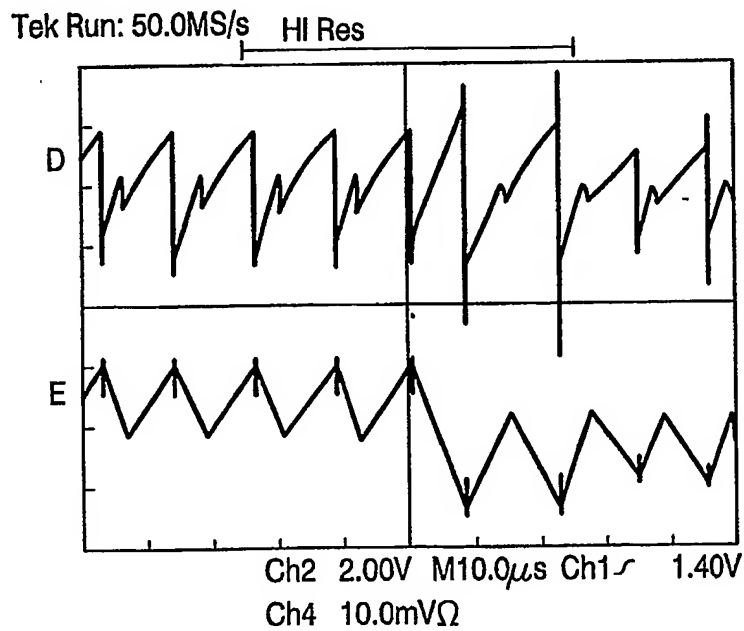
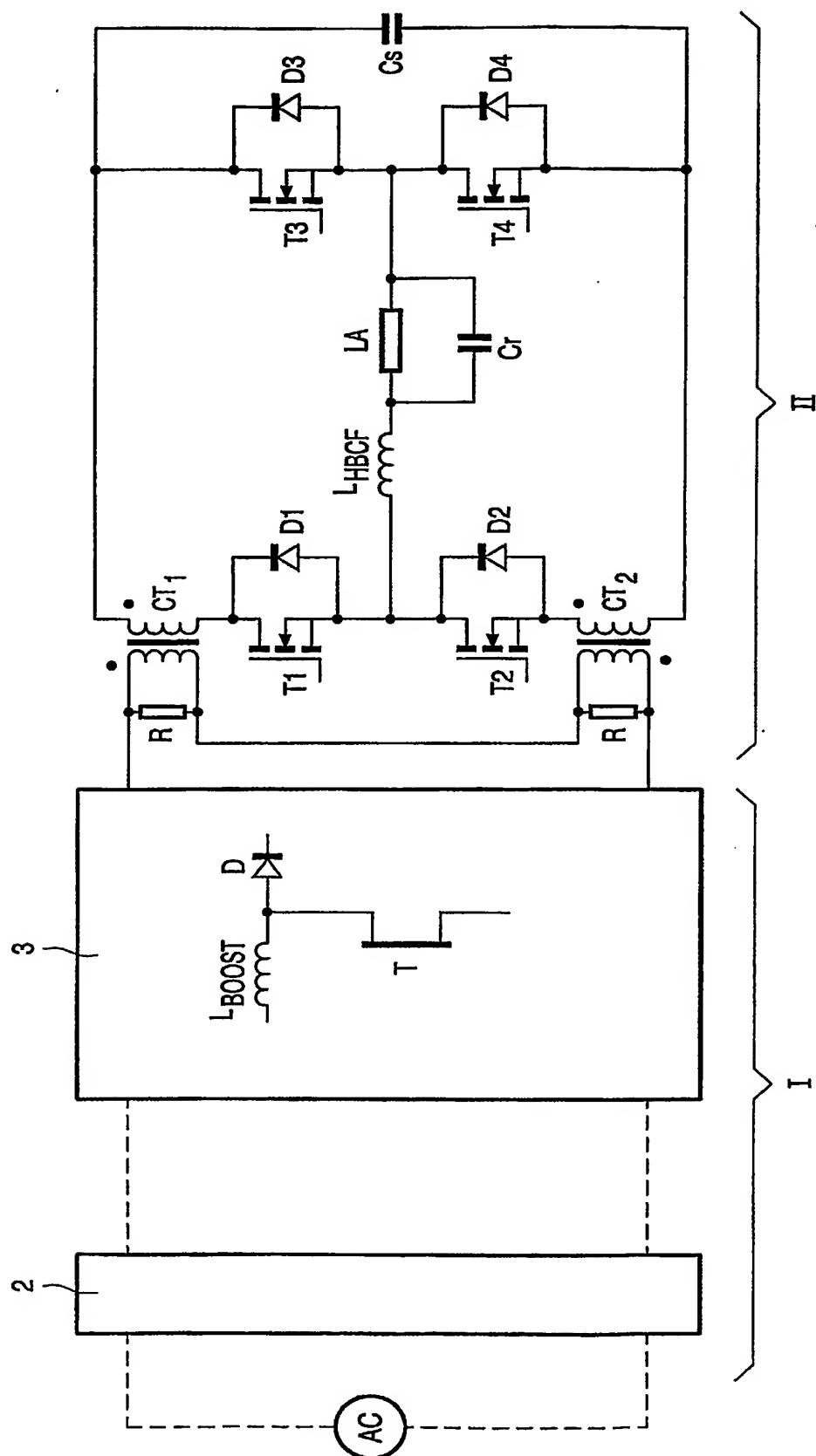


FIG. 3



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